Basics of Phased Arrays for Atmospheric and Geospace Science



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Phased Arrays

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Superposition Principle

Maxwell's Equations are Linear:

$$\begin{aligned} \mathbf{J}_{1} &= \frac{1}{\mu_{0}} \nabla \times (\mathbf{B}_{1}) - \epsilon_{0} \frac{\partial}{\partial t} (\mathbf{E}_{1}) \\ \mathbf{0} &= \nabla \times (\mathbf{E}_{1}) + \frac{\partial}{\partial t} (\mathbf{B}_{1}) \end{aligned}$$

$$\mathbf{J}_{2} = \frac{1}{\mu_{0}} \nabla \times (\mathbf{B}_{2}) - \epsilon_{0} \frac{\partial}{\partial t} (\mathbf{E}_{2})$$
$$0 = \nabla \times (\mathbf{E}_{2}) + \frac{\partial}{\partial t} (\mathbf{B}_{2})$$

$$\begin{split} \mathbf{J}_1 + \mathbf{J}_2 &= \frac{1}{\mu_0} \nabla \times (\mathbf{B}_1 + \mathbf{B}_2) - \epsilon_0 \frac{\partial}{\partial t} (\mathbf{E}_1 + \mathbf{E}_2) \\ 0 &= \nabla \times (\mathbf{E}_1 + \mathbf{E}_2) + \frac{\partial}{\partial t} (\mathbf{B}_1 + \mathbf{B}_2) \end{split}$$

Superposition Applied to Antenna Arrays



Fields radiated by single element at the origin with applied current I_0 :

$$\mathsf{E} = \mathsf{E}_0 l_0 rac{e^{-jk|\mathsf{r}|}}{|\mathsf{r}|}$$

Fields radiated by entire array:

$$\mathbf{E} = \mathbf{E}_0 \sum_{n=0}^{N-1} I_n \frac{e^{-jk|\mathbf{r}-\mathbf{r}_n|}}{|\mathbf{r}-\mathbf{r}_n|}$$

Far Field Approximation (Fraunhofer Zone)



If **r** and $\mathbf{r} - \mathbf{r}_n$ are almost parallel lines:

$$\mathbf{r} - \mathbf{r}_n \approx \mathbf{r} - |\mathbf{r}_n| \cos \theta \hat{r}$$

Assume $|\mathbf{r}_n| \ll |\mathbf{r}|$:

 $|\mathbf{r} - \mathbf{r}_n| \approx |\mathbf{r}| \text{ for demoninator terms}$ $-jk |\mathbf{r} - \mathbf{r}_n| \approx -jk |\mathbf{r}| + jk |\mathbf{r}_n| \cos \theta$ $\mathbf{E} \approx \underbrace{\mathbf{E}_0 \frac{e^{-jk|\mathbf{r}|}}{|\mathbf{r}|}}_{\text{Element Factor}} \underbrace{\sum_{n=0}^{N-1} I_n e^{jk|\mathbf{r}_n| \cos \theta}}_{\text{Array Factor}}$

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Simple Two Element Array Example

$$\frac{r}{\frac{d\cos\theta}{i_1} + \frac{d}{i_2}} z$$

Suppose $d = \lambda/4$ and

$$egin{aligned} \dot{i}_1(t) &= \cos\left(\omega t
ight) \ \dot{i}_2(t) &= \cos\left(\omega t + rac{\pi}{2}
ight) \end{aligned}$$

How does the radiated power vary as a function of θ ?

Two Element Array Example With Phasors



Active Electronically Steerable Phased Arrays

The AMISR UHF System





- 32 Antenna Element Units arranged in hexagonal pattern

- 3.5 x 2 meters; 19.8 dBi / panel
- 16 kW peak power per panel
- Basic system building block for AMISR
- Embedded linux controller



Panel (with PCU)

Utility Distribution Unit (UDU)

Face



AMISR Control System (ACS)

AMISR UDU

- 400 Hz JetPower converters
- Remote power control units
- Fiber distribution system

AMISR ACS

- Flexible transmit and receive system
- Completely remotely controlled
- Experiments run off a scheduler

Advanced Modular Incoherent Scatter Radar



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Ideal AMISR Radiation Pattern



AMISR Graceful Degradation



Steering Limits and Grating Lobes

- $d < \lambda/2 \rightarrow kd < \pi$: No grating lobes will ever appear
- $\lambda/2 < kd < \lambda \to \pi < kd < 2\pi$: Grating lobes will only appear at some steering angles
- $d > \lambda \rightarrow kd > 2\pi$: Grating lobes will always appear

Example of linear array with $d = 3\lambda/4$ spacing



• Maximum steering angle (for a linear array): $\Delta \theta_{\max} = |\sin^{-1} (1 - \frac{2\pi}{Ld})|$

AMISR Antenna Properties

- Hexagonal spacing with $d \approx 3\lambda/4$
- FOV limited by grating lobe limit $\sim 30^\circ 40^\circ$
- Antenna gain decreases with steering angle off of boresight
- Antenna works best within $\sim 25^{\circ}$ off of boresight



The PFISR Up-B Compromise



The Up-B beam is close to the grating lobe limit, and therefore has reduced sensitivity.

Reduced SNR in Up-B (Beam 2)



Conceptual Diagram of Steering with AMISR

