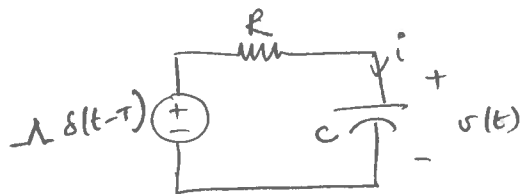
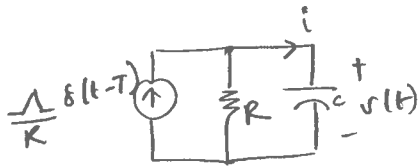


1 (a)



↓ Norton equivalent



The impulse current @  $t=T$  results into a charge dump on the capacitor which causes the voltage across the capacitor to change instantaneously:

$$\therefore v(T^+) = v(T^-) + \frac{\Lambda}{RC} = \boxed{V e^{-\frac{T}{RC}} + \frac{\Lambda}{RC} = v(T^+)}$$

$v(t^+)$  can also be found as follows.

KCL:

$$-\frac{\Lambda}{R} \delta(t-T) + \frac{v}{R} + C \frac{dv}{dt} = 0$$

Integrate both sides from  $t=T^-$  to  $t=T^+$

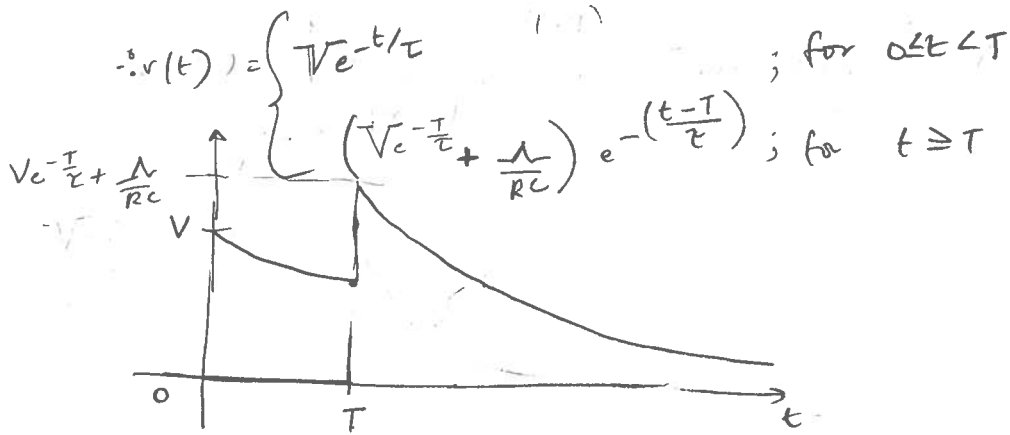
$$\int_{T^-}^{T^+} -\frac{\Lambda}{R} \delta(t-T) dt + \int_{T^-}^{T^+} \frac{v}{R} dt + C \int_{T^-}^{T^+} \frac{dv}{dt} dt = 0$$

$$\therefore -\frac{\Lambda}{R} + C [v(T^+) - v(T^-)] = 0$$

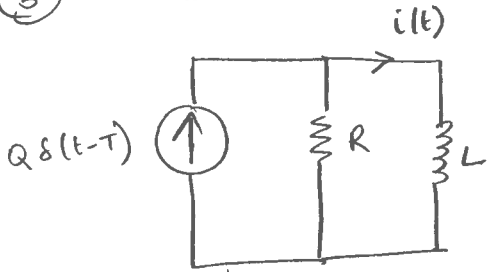
$$\therefore v(T^+) = C v(T^-) + \frac{\Lambda}{R}$$

$$\boxed{v(T^+) = V e^{-\frac{T}{RC}} + \frac{\Lambda}{RC}}$$

$v(\infty) = 0$  capacitor discharges through the resistor.



(b)



Given

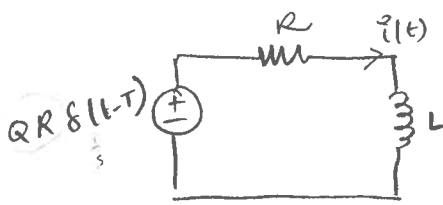
$i(0) = I$

$i(T^-) = I e^{-T/\tau} ; \tau = \frac{L}{R}$

$i(T^+) = ?$

$i(\infty) = ?$

Theremin equivalent



similar to that in part (a), the impulse voltage source results in a flux dump in the inductor causing the current to rise instantaneously

$$i(T^+) = i(T^-) + \frac{QR}{L}$$

$i(T^+)$  can also be found as follows.

KVL:  $QR \delta(t-T) - iR - L \frac{di}{dt} = 0$

Integrate both sides from  $t=T^-$  to  $t=T^+$

$$QR \int_{T^-}^{T^+} \delta(t-T) - R \int_{T^-}^{T^+} i dt - L \int_{T^-}^{T^+} \frac{di}{dt} dt = 0$$

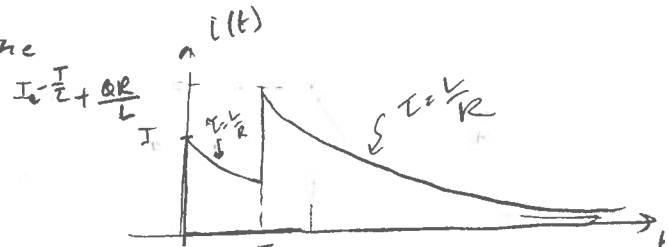
$QR - L [i(T^+) - i(T^-)] = 0$

$i(T^+) = i(T^-) + \frac{QR}{L}$   $i(T^+) = I e^{-T/\tau} + \frac{QR}{L}$

$i(\infty) = 0$  (Inductor discharges through the resistor)

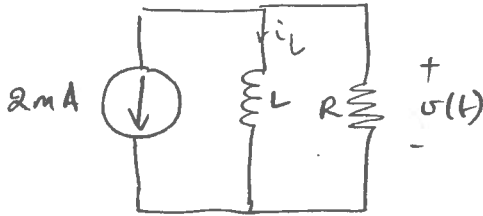
$$i(t) = I e^{-t/\tau} \quad 0 \leq t < T$$

$$= \left( I e^{-T/\tau} + \frac{QR}{L} \right) e^{-\frac{(t-T)}{\tau}} \quad \tau = \frac{L}{R}$$

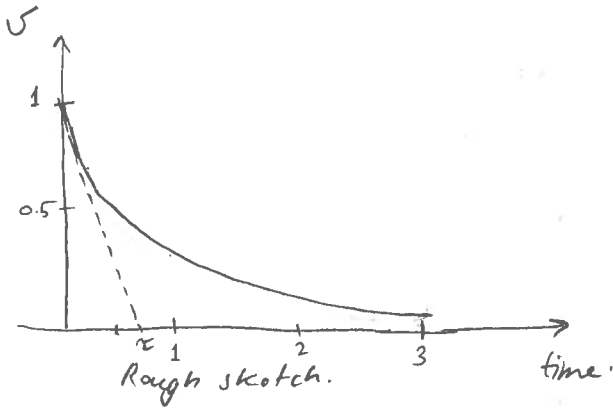


Note the duality between two circuits in part (a) of (b)

2

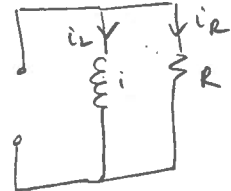


The measured  $v(t)$  is shown below.



Network had been assembled for a long time. At  $t=0$  the current source turns off.

At  $t=0^+$



$$i_L(0^+) = -2\text{mA}$$

$$i_L(0^+) = -2\text{mA}$$

$$v(0^+) = i_R(0^+) R$$

$$i_R(0^+) = -i_L(0^+) = 2\text{mA}$$

From the graph,

$$v(0^+) = 1\text{V} = 2\text{mA} \cdot R$$

$$\therefore R = \frac{1}{2} \text{ k}\Omega = \boxed{0.5 \text{ k}\Omega = R}$$

The time constant for the LC circuit is  $\frac{L}{R}$ . From the graph, the slope @ time  $t=0^+$  is  $\frac{1}{\tau} \approx \frac{1}{0.75}$ .

$$\therefore \tau = 0.75 \text{ sec}$$

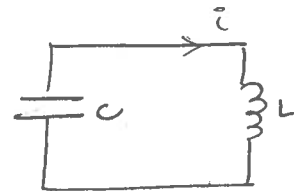
$$\therefore \frac{L}{R} = 0.75 \text{ sec}$$

$$\text{or, } L = 0.75 \times 0.5 \text{ k}\Omega \cdot \text{s} = 0.375 \text{ kH}$$

$$\boxed{L = 375 \text{ H}}$$

3 The natural frequency of the LC circuit is  $10^7 \text{ rad/s}$ .

$$\therefore \omega_0 = \frac{1}{\sqrt{LC}} = 10^7 \text{ rad/s} \Rightarrow C = \frac{1}{L(10^7)^2}$$



The peak value of  $v_C = 100 \text{ mV} = V_{\text{peak}}^C$

peak value of  $i_L = 20 \text{ mA} = I_{\text{peak}}^L$

The voltage across the capacitor reaches its peak value when  $i=0$ . and similarly  $i_L$  reaches peak when no energy is stored in the capacitor.

∴ By energy conservation,

$$\frac{1}{2} L I_{\text{peak}}^2 = \frac{1}{2} C V_{\text{peak}}^2$$

$$L (20 \text{ mA})^2 = \frac{1}{L (10^7)^2} (100 \text{ mV})^2$$

$$L^2 = \frac{(100 \text{ mV})^2}{(10^7)^2 (20 \text{ mA})^2} = 2.5 \times 10^{-13} \frac{\text{V}^2 \text{s}^2}{\text{A}^2}$$

$2.5 \times 10^{-6}$

$$\therefore L = 5 \times 10^{-7} \text{ H}$$

$$L = 0.5 \mu\text{H}$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$C = \frac{1}{L \omega_0^2}$$

$$C = \frac{1}{(10^7)^2 (0.5 \mu\text{H})} = 0.02 \mu\text{F}$$

(B) Total energy stored in the network =  $\frac{1}{2} C V_{\text{peak}}^2$

$$= \frac{1}{2} (1.86 \text{ mF}) (100 \text{ mV})^2$$

$$= 0.1 \text{ nJ}$$

[Check:  $\frac{1}{2} L I_{\text{peak}}^2$  should also yield the same answer]

$$= \frac{1}{2} (46.7 \text{ nH}) (20 \text{ mA})^2 = 0.1 \text{ nJ}$$

(C) Now a resistor is placed in parallel with the inductor and capacitor.

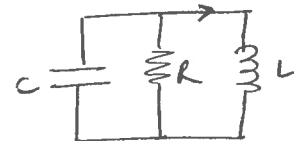
The energy is observed to decay

by the factor of  $\frac{1}{e}$  in 20 μs. ∴  $\tau = 20 \mu\text{s}$ .

For a parallel RLC circuit,

$$\frac{\partial^2 v_c}{\partial t^2} + \frac{1}{RC} \frac{dv_c}{dt} + \frac{1}{LC} v_c = 0$$

$$\alpha = \frac{1}{RC} \quad \omega_0 = \frac{1}{LC}$$

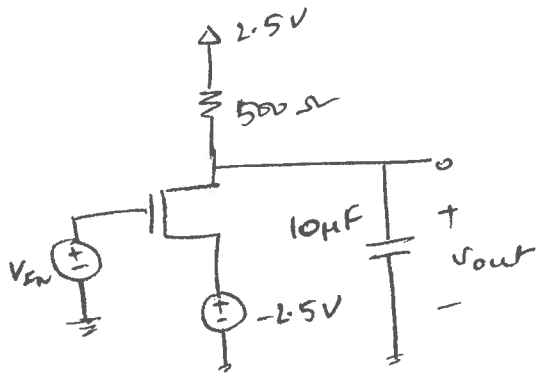


$e^{-t/\tau}$

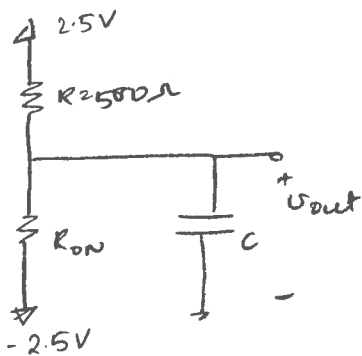
We know,  $\tau = \frac{1}{\alpha} = 2RC$

$$\therefore R = \frac{\tau}{2C} = \frac{20 \mu\text{s}}{2 \times 0.02 \mu\text{F}} = 500 \Omega = R$$

© Large signal model for MOSFET



For rising input



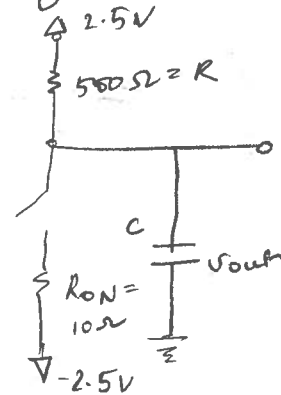
$$\therefore \tau = (R_{ON} \parallel R) C$$

$$= \frac{R_{ON} R}{R_{ON} + R} 10\mu F$$

$$\tau \approx 0.1 \text{ ms}$$

From the experimental data, it was hard to estimate  $\tau$  since the fall was really sharp. However it was clear that  $\tau \ll 1\text{ms}$  which agrees with the value calculated above.

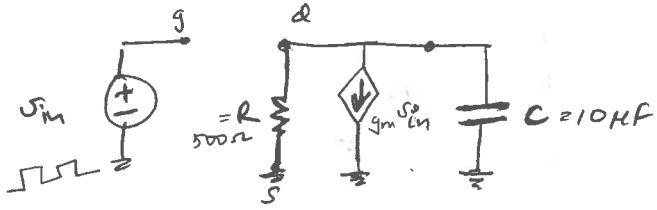
For falling input



$$\begin{aligned} \tau &= CR \\ &= 10\mu F + 500\Omega \\ &= 5 \times 10^{-3} \text{ sec} \\ \tau &= 5 \text{ ms} \end{aligned}$$

This matches the time constant acquired experimentally.

① Small signal model



$$R_{TH} = R = 500 \Omega$$

$$C = 10 \text{ nF}$$

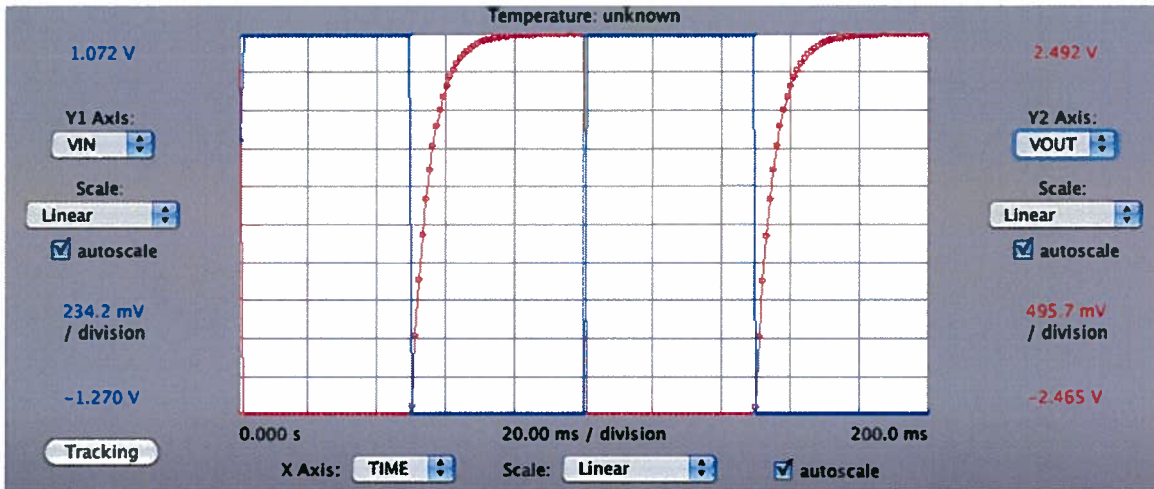
$$\therefore \tau = 5 \text{ msce}$$

The experimental data matches w/ the above value.

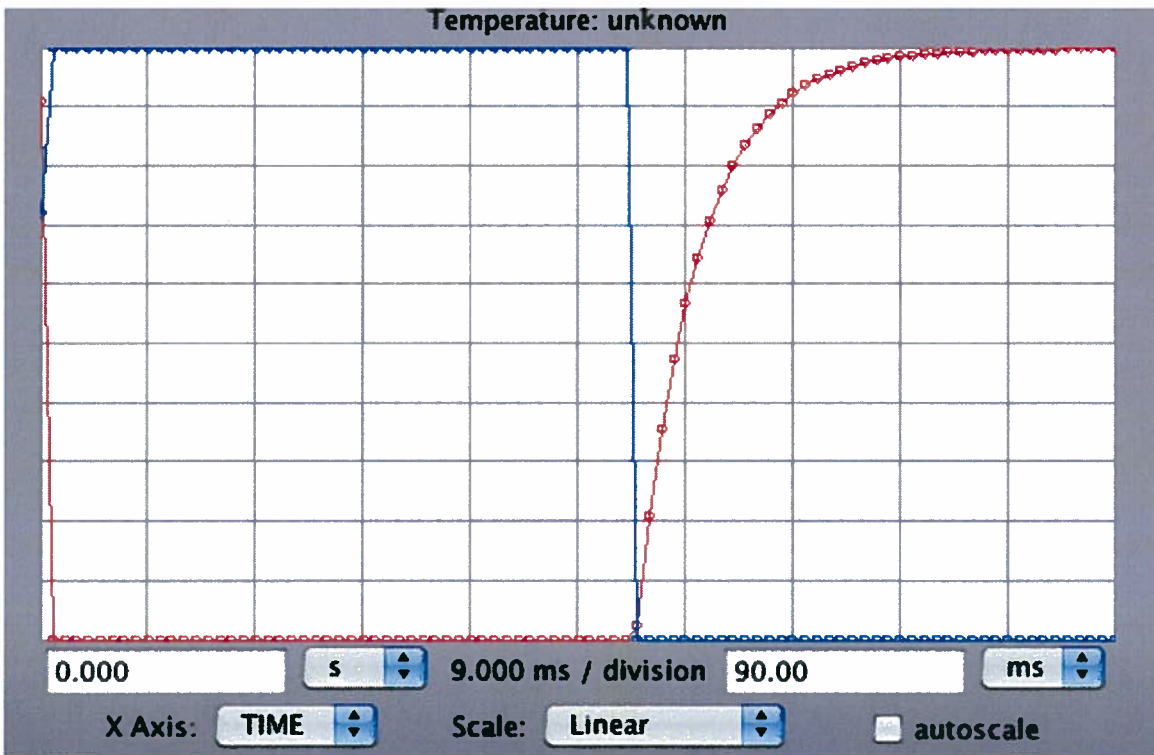
### Problem 4

Blue signal - Input  
Red signal - Output

A. For large signals



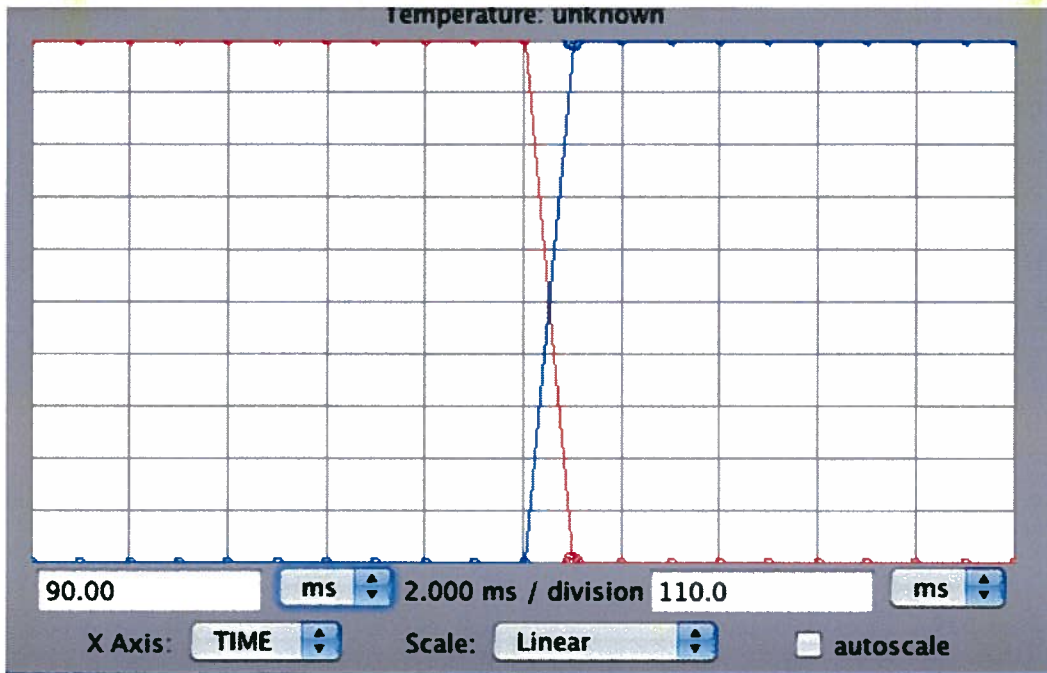
The rising and falling time constants can be found by finding the slope at the time when the output is rising and falling respectively.



The time constant when the output is rising is ~5 ms.

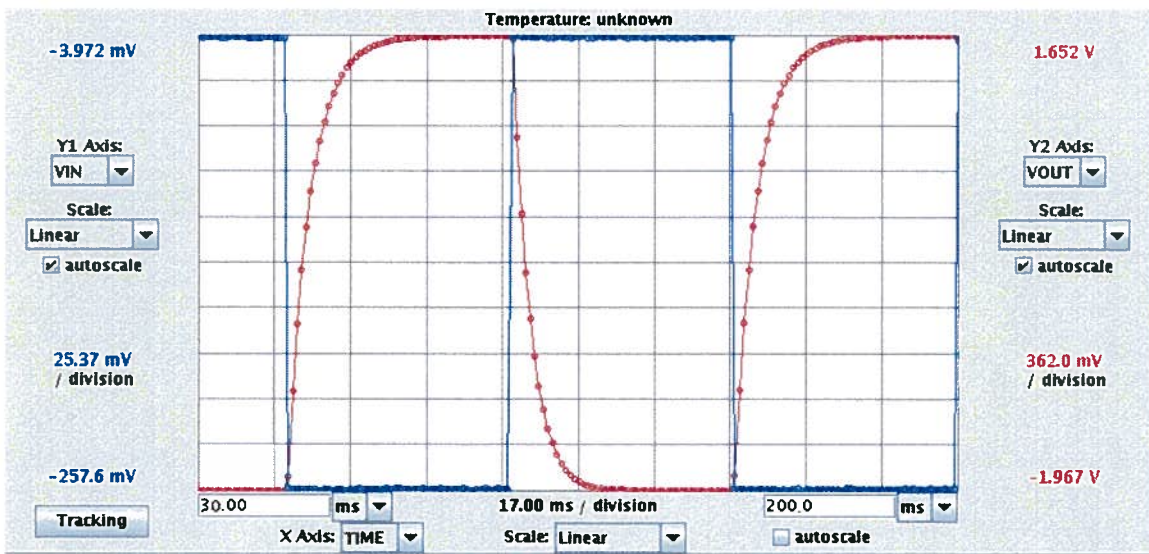






At falling edge of the output the time constant is really short as seen as a sharp fall in the output. From the above figure, it is seen that the time constant is much less than 1ms since the output falls from high to low in 1ms.

B . For small signals



The time constant for both rising and falling output is  $\sim 5$  ms.

