

Homework # 8 Solution

Due: Never

1. [16 Points]: Exercise 13.6

Answer:

(a) [4 Points] Plot (2)

It looks capacitive at low frequency, and resistive at high frequency.

(b) [4 Points] Plot (4)

$$H = [V_1(1/j\omega C)/(R+1/j\omega C) - V_1R/(R+1/j\omega C)]/V_1 \\ = (1-j\omega CR)/(1+j\omega CR) \\ \Rightarrow |H| = 1$$

(c) [4 Points] Plot (8)

At low (or high) frequency, the capacitor (or inductor) looks like an open circuit. Therefore,

$$Z(\omega=0) = Z(\omega=\infty) = R$$

At resonance, the LC in series looks like a short circuit. That is,

$$Z(\omega=\omega_0) = 0$$

(d) [4 Points] Plot (5)

At low frequency, the inductor looks like a short circuit. Therefore,

$$Z(\omega=0) = R_1$$

At high frequency, the inductor looks like an open circuit. Therefore,

$$Z(\omega=\infty) = R_1 + R_2$$

In between, it's a first order system, so the slope of $Z(j\omega)$ is 1.

2. [20 Points]: Problem 13.4

Answer:

(a) [5 Points]:

$$H = \frac{R_1 + j\omega L_1}{R_1 + R + j\omega(L_1 + L)}$$

(b) [20 Points]:

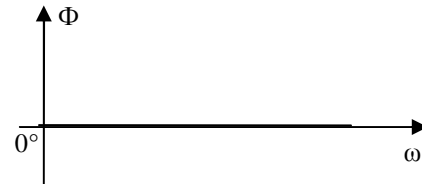
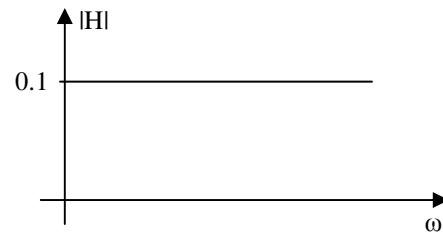
$$H(\omega=0) = R_1/(R_1+R) = 1/10 \\ \mathbf{R = 9k\Omega}$$

(c) [5 Points]:

$$H(\omega=\infty) = L_1/(L_1+L) = 1/10 \\ \mathbf{L = 90mH}$$

(d) [5 Points]:

$$H(j\omega) = 1/10$$



3. [24 Points]: Problem 13.4

Answer:

(a) [5 Points]:

$$i(t) = I \cos(\omega t) \\ i = I e^{j\omega t}$$

$$v = Zi = \frac{i}{\frac{1}{j\omega L} + j\omega C + \frac{1}{R}} \\ = \frac{i}{j\left(\omega C - \frac{1}{\omega L}\right) + \frac{1}{R}}$$

Therefore,

$$v(t) = V \cos(\omega t + \Phi)$$

$$V = \frac{I}{\sqrt{\frac{1}{R^2} + \left(\omega C - \frac{1}{\omega L}\right)^2}}$$

$$\Phi = \tan^{-1}\left(R\left(\frac{1}{\omega L} - \omega C\right)\right)$$

(b) [4 Points]:

$$\text{At } \omega = \frac{1}{\sqrt{LC}},$$

$$V = V_{MAX} = IR$$

(c) [5 Points]:

$$\omega = \frac{1}{\sqrt{LC}},$$

$$C = \frac{1}{L\omega^2}$$

$$\begin{aligned} C_1 &= 1/L\omega_1^2 \\ &= 1/(365\mu \cdot (2 \cdot 3.14 \cdot 540k)^2) \\ &= 2.38 \cdot 10^{-10} \text{ C} \\ &= 238 \text{ pF} \end{aligned}$$

$$\begin{aligned} C_2 &= 1/L\omega_2^2 \\ &= 1/(365\mu \cdot (2 \cdot 3.14 \cdot 1600k)^2) \\ &= 2.71 \cdot 10^{-11} \text{ C} \\ &= 27.1 \text{ pF} \end{aligned}$$

Therefore, C must be able to vary from **27.1pF to 238pF**.

(d) [5 Points]:

$$V = \frac{IR}{\sqrt{1 + \left(\omega C - \frac{1}{\omega L}\right)^2 R^2}} = 0.25IR$$

$$1 + \left(\omega C - \frac{1}{\omega L}\right)^2 R^2 = 16$$

$$\begin{aligned} R &= \sqrt{15} \left(\omega C - \frac{1}{\omega L}\right)^{-1} \\ &= \sqrt{15} \left((\omega_0 + \Delta\omega)C - \frac{1}{(\omega_0 + \Delta\omega)L} \right)^{-1} \end{aligned}$$

$$= \sqrt{15} \left(\Delta\omega C + \frac{\Delta\omega}{\omega_0^2 L} \right)^{-1}$$

Notice $C = \frac{1}{L\omega_0^2},$

$$\begin{aligned} R &= \sqrt{15} \left(\frac{2\Delta\omega}{\omega_0^2 L} \right)^{-1} = \frac{\sqrt{15}\omega_0^2 L}{2\Delta\omega} \\ &= \frac{\sqrt{15} \times (6.28 \times 1M)^2 \times 365\mu}{2 \times 6.28 \times 5k} \\ &= 888k\Omega \end{aligned}$$

(e) [5 Points]:

$$\frac{V}{V_{MAX}} = \frac{1}{\sqrt{1 + \left(\omega C - \frac{1}{\omega L}\right)^2 R^2}}$$

$$\approx \frac{1}{\left(\omega C - \frac{1}{\omega L}\right)R} \approx \frac{\omega_0^2 L}{2\Delta\omega R}$$

$$\begin{aligned} &= \frac{(6.28 \times 1M)^2 \times 365\mu}{2 \times 6.28 \times 10k \times 888k} \\ &= 0.13 \end{aligned}$$

$$Q = \frac{R}{\omega_0 L} = \frac{888k}{6.28 \times 1M \times 365\mu} = 389$$

4. [40 Points]:

Answer:

(a) [5 Points]:

See the figure on the next page.

(b) [6 Points]:

$$T = 150\mu\text{s}$$

Frequency:

$$f = 1/T = \mathbf{6.67 \text{ kHz}}$$

Time constant:

$$\tau = 1/\alpha = \mathbf{300 \mu\text{s}}$$

Amplitude at $t=0^+$:

$$V = \mathbf{330 \text{ mV}}$$

These values are similar at the rising edge and falling edge of the square wave.

(c) [5 Points]:

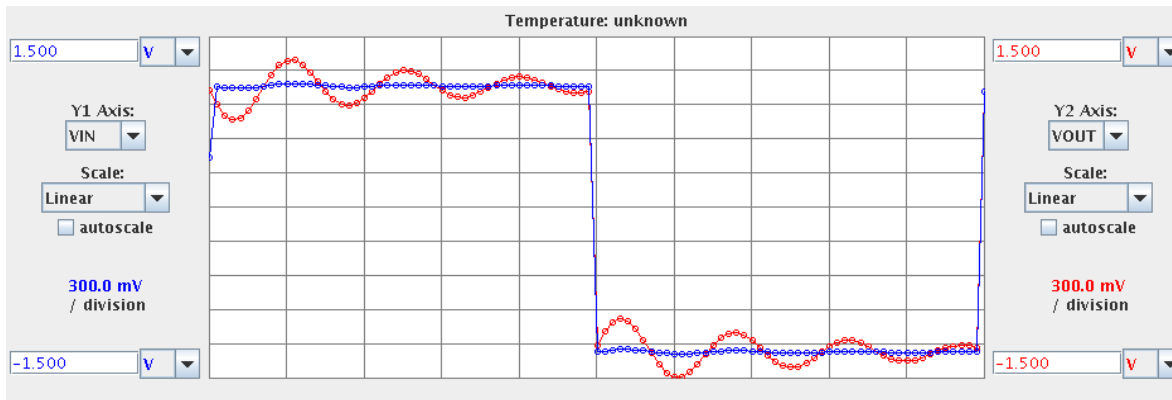
$$v_{IN} = v_R + v_C + v_L$$

Differentiate,

$$\frac{dv_{IN}}{dt} = R \frac{di}{dt} + \frac{1}{C} i + L \frac{d^2 i}{dt^2}$$

(d) [5 Points]:

	v_C	v_L	v_R	v_{OUT}	i	di/dt
$t=0^-$	V_{IN1}	0	0	V_{IN1}	0	0
$T=0^+$	V_{IN1}	$V_{IN2} - V_{IN1}$	0	V_{IN2}	0	$(V_{IN2} - V_{IN1})/L$



(e) [8 Points]:

In steady state, $v_C = \text{constant}$, therefore,
 $i(\infty) = 0$,

because there is no DC current through the capacitor.

The general solution to a second order differential equation is a linear combination of two exponential functions, and if we know it's oscillating, then it must have the following form,

$$i(t) = I \sin(\omega t + \Phi) e^{-\alpha t}$$

$$\frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = \frac{1}{L} \frac{dv_{IN}}{dt}$$

1) $i(t=0+) = 0$, therefore,
 $\Phi = 0$

$$2) \omega = \sqrt{\frac{1}{LC} - \left(\frac{R}{2L}\right)^2}$$

3) $\alpha = R/2L$

$$4) \left. \frac{di}{dt} \right|_{t=0+} = \omega I = \frac{V_{IN2} - V_{IN1}}{L}$$

$$I = \frac{V_{IN2} - V_{IN1}}{\sqrt{\frac{L}{C} - \frac{R^2}{4}}}$$

(f) [6 Points]:

$$v_{OUT} = v_{IN} - v_R$$

$$= v_{IN2} - R I \sin(\omega t + \Phi) e^{-\alpha t}$$

For the rising edge,

$$v_{IN1} = -1V, v_{IN2} = 1V \text{ (nominal)}$$

For the falling edge,

$$v_{IN1} = 1V, v_{IN2} = -1V \text{ (nominal)}$$

I , ω , Φ , and α are calculated in part (e). Remember, I is also a function of v_{IN1} and v_{IN2} .

(g) [5 Points]:

From part (b),

$$RI = \frac{R(V_{IN2} - V_{IN1})}{\omega L} = 0.33V$$

From the graph above,

$$v_{IN2} - v_{IN1} = 2.3V$$

Therefore,

$$\omega L/R = 2.3/0.33 = 6.9$$

$$L/R = 1.6 \cdot 10^{-4} \text{ sec}$$

which agrees with the measurement

$$1/\alpha = 2L/R = 300 \text{ us}$$

$$L/R = 1.5 \cdot 10^{-4} \text{ sec}$$

Substitute this into the equation of ω ,

$$1/LC = \omega^2 + (R/2L)^2 = 2.9 \cdot 10^8 \text{ Hz}^2$$

$$LC = 2.5 \cdot 10^{-9} \text{ sec}^2$$

However, we are not able to figure out the resistor, capacitor, and inductor values from the measurements of voltage transfer function (VTF). As long as L/R and LC have the values specified above, we will get the same VTF no matter what absolute values of R , L , and C we have chosen.